

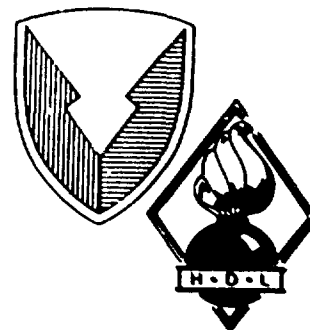
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May 1989

HEMP Validation of FAA Radio Facility

by Stephen C. Sanders

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U.S. Army Laboratory Command
Harry Diamond Laboratories
Adelphi, MD 20783-1197

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<p>The Harry Diamond Laboratories (HDL) provided high-altitude electromagnetic pulse (HEMP) and partial lightning hardness verification testing to a high-frequency radio facility for the Federal Aviation Administration (FAA), under the Department of Transportation (DOT). The facility consisted of a shielded enclosure with antennas and cables connected through protective devices at the shield to dummy loads within the shield. It was exposed to low-level radiation with measurements of internal fields extrapolated to threat levels and compared to specific criteria. Each typical penetration was injected with a simulated lightning source, and internal measurements were compared to specified criteria. Although criteria were not always met, it was judged with confidence that the facility would protect radio equipment. It was noted that the ac power spark gap would remain shorted until ac power shut down, resulting in system upset.</p>					
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1. INTRODUCTION

1.1 Program

The Federal Aviation Administration (FAA) is preparing a high-frequency radio facility (HFRF) that is resistant to both high-altitude electromagnetic pulse (HEMP) and normal lightning to communicate between air traffic and controllers. This HFRF has a shielded enclosure which was tested for its ability to provide protection from the effects of HEMP and lightning. This test report describes the tests performed on the HFRF by the Systems Management American Corporation (SMA) and Harry Diamond Laboratories (HDL).

1.2 Objectives and Criteria

The primary objective of the testing was to verify beyond reasonable doubt that the equipment within the HFRF will be able to perform its mission after a HEMP or a normal lightning event with no loss of communications links or situation data. A secondary objective was to substantiate threat responses and establish a high level of confidence in the validity of the test. As part of these objectives, hardening devices were tested to establish their performance margins. The HEMP threat criterion is provided by the quadripartite standard treaty agreement group document, QSTAG-620, supplemented with an injection criterion to simulate the coupled responses of long cables; simulation is necessary because it is impractical to duplicate the threat environment over long distances. The lightning criterion, provided by the FAA, is tested by current injection to demonstrate that the equipment would survive most lightning strokes. The shield-room criterion calls for 60 dB of attenuation over the frequency range from 10 kHz to 1 GHz. Penetrations of this shield must meet a

transient limiting criterion of 100 V, 10 A, except for the ac power (110/220 V), specified as 1 kV, 10 A, and the two rf coaxial penetrators, specified as 10 kV, 500 A.

1.3 Site Description

The facility tested consisted of a transportable shielded room assembled at the Repetitive EMP Simulator (REPS) site at HDL's Woodbridge Research Facility. The shielded room was penetrated by ac and dc power lines, control lines, telephone lines, and rf lines elevated parallel to the simulator. All lines were shielded except the ac power lines. The room was designed to contain rack-mounted hf radio equipment, amplitude modulated, single sideband, receiving or transmitting over the frequency range of 2 to 29.999 MHz. The rf lines were terminated in the tuner circuitry in the shelter and with a 35-ft dipole at the far end. Other lines were terminated at both ends with resistors. Table 1 lists the circuits of interest for HEMP evaluation.

1.4 Threat Description

A high-altitude nuclear explosion generates a large amount of electromagnetic energy in the ionosphere which radiates about 1 J/m² over hundreds of thousands of square miles of the earth's surface. This energy can couple into metallic loops or lengths which can then conduct through an electrical unit's interface to devices which may be unable to withstand the stress. Loss of logic or operating characteristics may result. Some thresholds of damage are as follows:

relays	1 to 10 J
transistors	1 to 100 mJ
integrated circuits	10 to 1000 μ J

Table 1. HEMP penetration points

Function	Penetrations	Maximum cable length (ft)	Cable type	Protection	Load
ac power	2 sets, 3 wire	300	Unshielded	400-V spark	4.7 Ω
Audio	39 pair, 2 wire	300	Shielded	12-V Zener	650 Ω
Digital	4 pair, 2 wire	300	Shielded	12-V Zener	120 Ω
Phones	2 sets, 4 wire	300	Shielded	100-V Zener	650 Ω
rf line	2 sets, coax	270	Coaxial	2000-V spark	Complex*
Receive lines	1 line, coax	150	Unshielded	250-V spark	50 Ω

*16-MHz tuned circuit

A thunderhead builds an electrical charge until a "streamer" of charge advances along a path of dielectric breakdown, or ionization, while an opposite streamer of charge rises from the earth to meet it. When they join, a current (of from 3 to 100 kA) drains the charge. As the charge depletes, other nearby charges resupply it for followup currents, or strokes. These strokes prefer conductors as a target and are destructive to exposed electrical equipment. Table 2 describes various lightning threats as a function of probability.

Protection from both threats requires that the transient energy be isolated from the equipment, usually by being diverted to an earth ground. These tests verify that the protection works.

1.5 Test Facilities

A current injector of 20 kV, 200 J of energy and a source impedance of 6 Ω was used to simulate lightning on the cables. This injector consists of a bank of capacitors charged until a spark gap pressurized with sulfur hexafluoride (SF_6) arcs, at which time the charge flows through a pulse-

shaping network to the circuit under test, returning to the injector through facility ground. HDL's REPS, which can radiate a low-level composite HEMP over a 50 by 200 m area, provided typical cable response data. REPS provides a 1.5-kV/m horizontal, ground-interacted electric field at the target, 100 m on the centerline, 50 m off the centerline, and 3 m high, with threat-related properties. REPS consists of a Marx generator driving a 1000-ft biconic antenna elevated 60 ft.

Instrumentation was provided by the Mobile Digital Acquisition System (MODAS). This van includes three channels of fiber-optic data linkup to the transmitter and probe, two digitizers per channel, data reduction equipment, and displays. Table 3 describes the limitations of the instrumentation used in the test.

Personnel consisted of an electronics technician, a simulator operator, and a test engineer. Tests were conducted daily from 21 April through 8 May 1987, began at 8:00 am, and ended after checkout and shutdown at 4:00 pm.

Table 2. EMP and lightning parameters

Parameter	All lightning strikes			HEMP ≤50 kV/m
	5% <1 kA	50% <3.3 kA	95% <100 kA	
Risetime 10-90%	0.28 μs	0.8 μs	8 μs	0.01 μs
Falltime to 50%	45 μs	130 μs	170 μs	0.4 μs
Slew rate	3.6 kA/ μs	4.3 kA/ μs	12.5 kA/ μs	5000 kV/ μs

Source: Cyanos and Pierce, *A Ground-Lightning Environment*, Tech Report 1, Stanford Research Institute (August 1972).

Table 3. Instrumentation characteristics

Instrument	Bandwidth	Error (%)
SRI E010 and E204 E-field sensors	10 kHz to 10 MHz	10
Tektronix P6009 voltage probe	dc to 400 MHz	2
Tektronix P6021 current probe	100 Hz to 200 MHz	10
Stoddard 94430 current probe	20 Hz to 200 MHz	15
Nanofast Optic transmitter/receiver set	dc to 120 MHz	5
Tektronix digitizer 7912AD	dc to 200 MHz	5

2. HEMP RADIATION TESTS

2.1 Test Setup

Before testing, the shielded room was erected upon a stable wood platform about 80 m from the REPS and 50 m off the centerline, as shown by figure 1. Instruments for mapping the outside and inside of the shielded room and for measuring voltages and currents each side of the penetration protective devices were set up and calibrated. The test points are described in table 4.

Tests were performed on two of each type of line so as to sample two of each protective device (e.g., AUD1 and AUD2). Voltage measurements were single ended to ground; inner conduc-

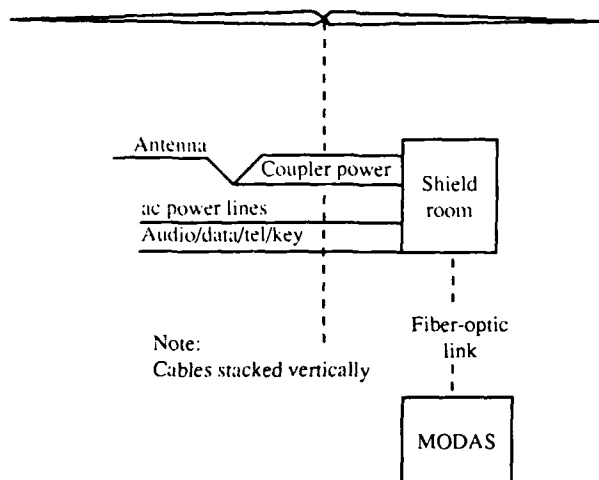


Figure 1. Radiated HEMP test configuration and test points.

Table 4. Test point description

At all test points except mapping points EXO and EXI, outside voltage, inside voltage, and inside current were measured.

Test point	Location
EXO	Mapping point 1 m in front, 1 m from ground
EXI	Mapping point 1 m inside, 1 m from floor
SAC	400-V (4.7-Ω) wire at entry panel
SACN	200-V (15-Ω) wire at entry panel
COU	39-V (120-Ω) wire at entry panel
AUD	12-V (650-Ω) wire at entry panel
DAT	12-V (160-Ω) wire at entry panel
TEL	110-V (650-Ω) wire at entry panel
KEY	22-V (1-kΩ) wire at entry panel
XMT	2000-V (tuner) coax at tuner unit
RCV	250-V (50-Ω) coax at load resistor

tor current measurements were taken from around the same conductor. The pulse injector was set up outside the entry panel. An E-field mapping box was centrally placed, 1 m in front of the shielded room, at a 1-m elevation.

2.2 Test Procedure for Pulsed E-Field Measurements

1. Align the E-sensor on the mapping box to measure the field parallel to the REPS antenna. Verify the settings for expected trigger, amplitude, risetime, and ringdown.
2. End-to-end calibrate the instrumentation and record a pulse.
3. Scale the instrumentation and digitizer for resolution.
4. Pulse the test point as often as necessary to record data with repeatability and reversibility. On occasion, record the level of ambient noise and instrumentation-coupled HEMP.
5. Evaluate the data.
6. Move the mapping box into the shielded room 1 m from the center of the front wall, and 1 m above the floor. Align the sensor exactly as was done for the outside measurements. Repeat steps 3, 4, and 5 for test points EXI, EYI, and EZI.
7. At the conclusion of each test period, turn off the instruments, secure the area, and notify the REPS personnel.

2.3 Test Procedure for cw E-Field Measurements (MIL-STD-285)

1. Set up transmitter, receiver, and antenna outside the HFRF. Connect feedwires to the HFRF at test points 1 and 23.
2. Turn on the transmitter and tune to 95 kHz.
3. Insert the receiving antenna into the calibration port of the transmitter and adjust the gain control until 0 dB is read on the receiver meter.

4. Move the receiving antenna to each of the 12 different positions inside the RFI-tight HFRF and record the decibel meter reading.

5. Repeat steps 2 through 4 using 450 MHz and the Retlif TS450 test set and calibrating procedures, about 10 ft from the HFRF.

2.4 Test Procedure for Coupled Responses

1. Attach the voltage and current probe to the chosen 120-V line inside the entry panel. Attach a voltage probe to the same line outside the entry panel. Verify the settings for the expected response.

2. End-to-end calibrate the instrumentation and record a pulse.

3. Scale the instruments and digitizer for resolution.

4. Pulse the test point as often as necessary to record data with repeatability and reversibility. On occasion, record the level of ambient noise and instrumentation-coupled HEMP.

5. Evaluate the data.

6. Proceed to the next test point and repeat steps 3, 4, and 5.

7. At the conclusion of each test period, turn off the instruments, secure the area, and notify the REPS personnel.

2.5 Test Procedure for Injection Responses

1. Attach the voltage and current probe to the chosen ac line inside the entry panel. Attach a voltage probe to the same line outside the entry panel. Verify the settings for expected triggering, amplitude, risetime, and ringdown.

2. End-to-end calibrate the instruments and record a pulse.

3. Scale the instruments and digitizer for optimum resolution of amplitude and time.

4. Pulse the test point as often as necessary to record data with repeatability and reversibility.

5. Evaluate the data.

6. Proceed to the next test point and repeat steps 3, 4, and 5.

7. At the conclusion of each test period, turn off the instruments and secure the area.

3. TEST DATA REDUCTION

Figures 2 through 13 show test data; all data were recorded at least twice to demonstrate repeatability and occasionally to show reversed polarity.

3.1 Electrical Field

Figure 2 shows the electrical fields with and without REPS. An EMP free field at threat level is 50 kV/m within 10 to 25 ns of start. The reflection of this free field off the electrical ground causes cancellation with a delay depending on the height of measurement. Vertical fields and radial fields are of minor strength, so the shielding effectiveness of the HFRF can be reasonably accurately represented as

$$SE = 20 \log(EXI/EXO) .$$

The 1-m-high data show 220 V/m after 13 ns, because of cancellation of the field before the free field maximum arrives. The same location with a 3-m height shows 1.6 kV/m after 25 ns, demonstrating a time for the peak to arrive before the ground cancelling wave arrives. This 1.6-kV/m value is comparable to a threat amplitude. The threat scale factor is therefore (50 kV/m)/(1.6 kV/m) or 0.032. The shielding effectiveness is calculated as

$$\begin{aligned} SE &= 20 \log(0.04/150) \\ &= -72 \text{ dB} . \end{aligned}$$

This value is better than the -60 dB criterion. The EMP level inside the HFRF resembles the noise level outside.

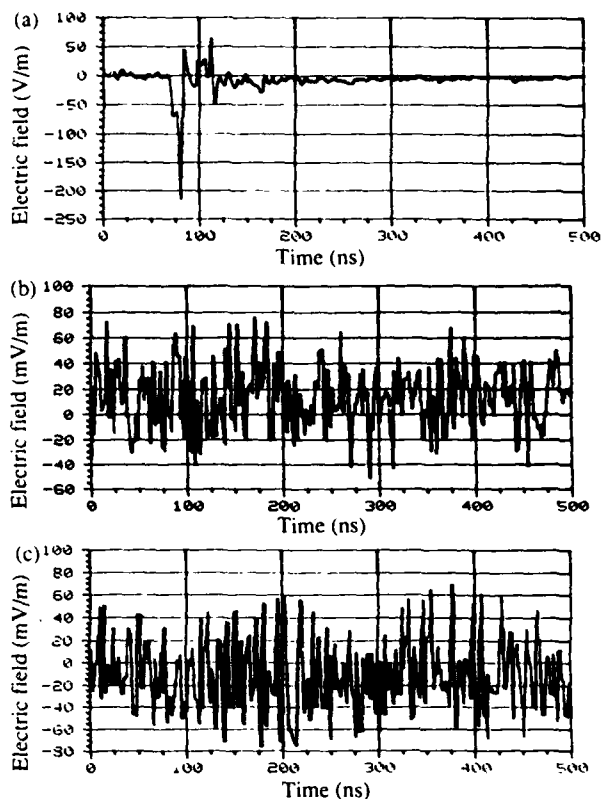


Figure 2. Electric field data: (a) horizontal field outside shelter with REPS on, (b) ambient horizontal field outside shelter (without REPS), and (c) horizontal field inside shelter (with REPS).

3.2 cw field

Figure 3 shows the response of the HFRF to the cw field. Although the shielding effectiveness varied day to day, it always exceeded the criterion of -60 dB. The HFRF was not intended to be exposed to the extremes of temperature and humidity that it underwent during this testing (it sat on a wood platform under a tent outdoors during April), but the use of brass wool under the bolts and the lack of calibrated torquing for evenly distributed seam bonding pressure suggest that the HFRF shielding effectiveness could be substantially improved. The shield passes the criterion.

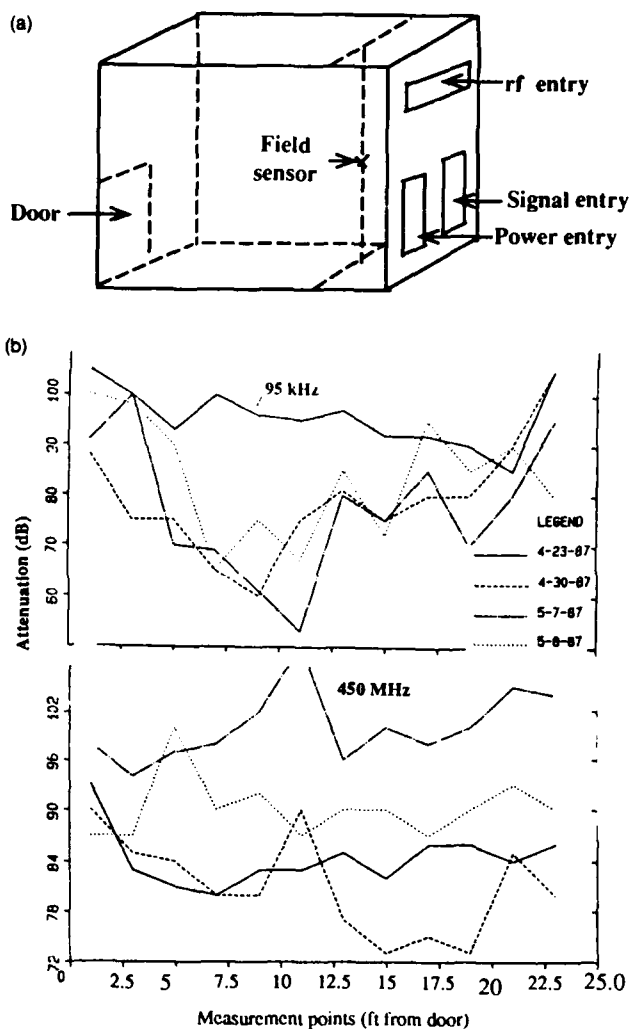


Figure 3. Response of HFRF to cw field: (a) field sensor location in shield room and (b) field measurements over time.

3.3 ac Power Lines

The response of the ac power source to REPS (fig. 4) shows a 10-V, 0.5-A (20- Ω) penetration to the 4.7- Ω load. When scaled to threat, this represents 320 V, 16 A, still below the protection level of 400 V. The 0.7-M' \angle ringing observed represents the 300-ft line quarter-wave resonance:

$$f = 0.8c/\lambda,$$

where f = frequency,
 c = speed of light, and
 λ = wavelength.

Since the current travels 300 ft (which equals a quarter wavelength), the frequency = 0.667 MHz. The injection voltage shows a 17-MHz ringing. Since the cables are not connected, this ringing represents the reflection of the injection current at the impedance mismatch of the shorting protective spark gap. Previous predictions, based on an in-line filter, are invalid. The injected pulse measured 400 V, 14 A (28 Ω), equal to the 400 V of the protective device. The 4.7- Ω load looks like 20 to 30 Ω to the transient.

The ac power source lines are judged to be hard to EMP, and to half of all probable lightning strokes as well, although the current criterion of 10 A was exceeded. A 1-kV pulse on an apparent 20- Ω load would yield 50 A, which would be a more appropriate criterion to use.

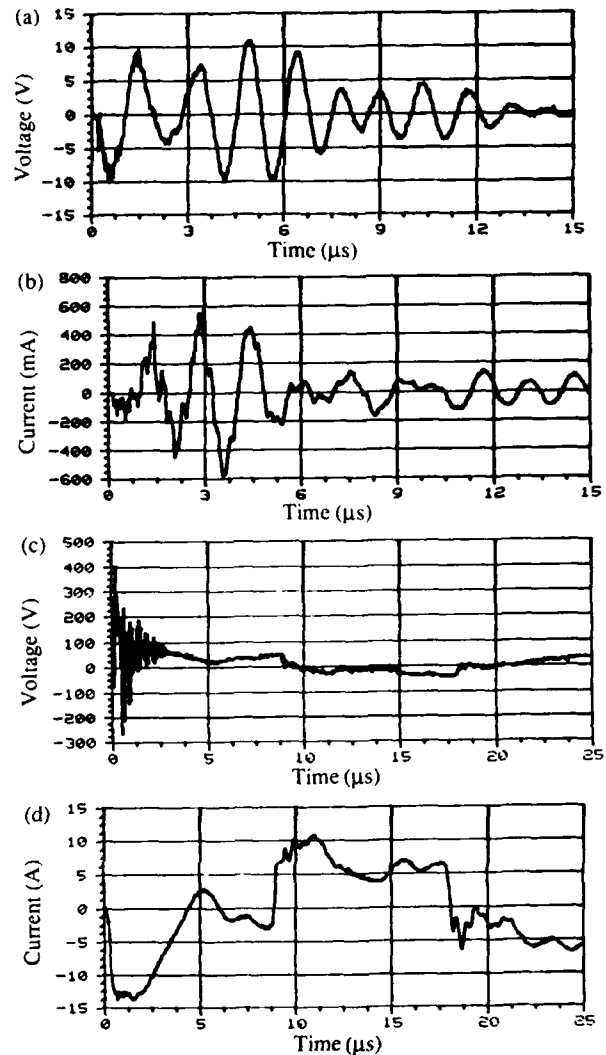


Figure 4. Response of ac power source line: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to injection, and (d) current response to injection.

3.4 ac Power Neutral Line

The response of the ac power neutral to REPS (fig. 5) shows a 7-V, 0.5-A (14- Ω) penetration to the 4.7- Ω load. When scaled to the threat, this represents 250 V, 16 A, just above the protection level of 200 V. The 0.7-MHz ring observed represents the 300-ft line resonance.

Injection data show a 0.11-MHz ringing. The cables are not connected, so this represents the resonance of the injection reactance and the HFRF reactance. Data show 60 V, 14 A (4.2 Ω), exceeding the 400 V of the protective device. No explanation of this low and inconsistent apparent resistive value can be found, so the data are considered bad. The ac power source neutral lines are hard to EMP. The protective device found to be effective on the phase line should be twice as effective on the neutral line, so this line is assumed (but not proven) to be hard to half of all lightning strokes as well.

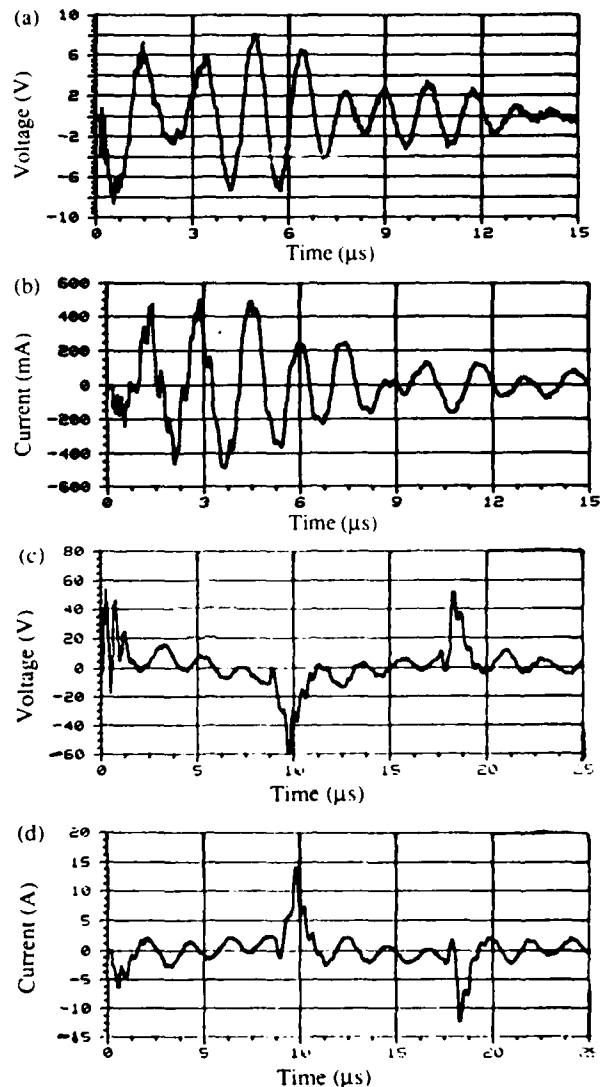


Figure 5. Response of ac power neutral line: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to injection, and (d) current response to injection.

3.5 ac Power Auxilliary Line

The response of the ac power auxiliary line to the REPS field (fig. 6) shows a 12-V, 0.5-A (24- Ω) penetration to the 15- Ω load. When scaled to threat, this represents 375 V, 16 A, still below the protection level of 400 V. The 0.7-MHz ring observed represents the 300-ft line resonance.

Injection data show a 0.11-MHz ringing. The 120-V, 15-A response (8 Ω) is less than the 200 V of the protective device. Because the virtual impedance measures less than that for the real load with no cable, we conclude that these data are bad.

The ac power auxiliary lines are hard to EMP. The protective device found effective on the source phase line will probably (and should) be installed on the auxiliary lines, so this line is assumed (but not proven) to be hard to half of all lightning strokes as well.

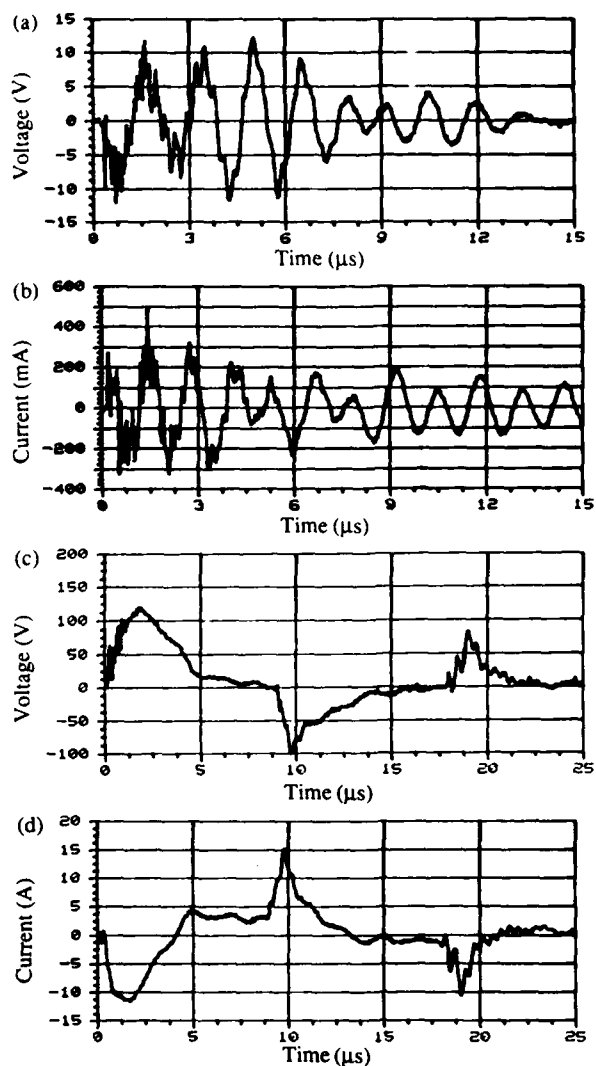


Figure 6. Response of ac power auxiliary line: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to injection, and (d) current response to injection.

3.6 ac Power Coupler

The response of the ac power coupler to the REPS field (fig. 7) shows a 330-mV, 20-mA (17- Ω) penetration to the 120- Ω load (see below). These low values, when compared to the unshielded power lines (10 V), show an effective cable shield of 30 dB. When scaled to threat, this represents 10 V, 6 A, still below the protection level of 39 V. The 0.7-MHz ringing observed represents the 300-ft line quarter-wave resonance. The shielded cable characteristic impedance dominates the transient, rather than the load (which is suspected to have been 12 Ω , not 120 Ω).

Injection data show a 0.11-MHz ringing. The data show a 300-V, 14-A (22- Ω) response into 120 Ω ; this response rings as if there were a spark gap mismatch around 200 V, 7 A, decaying to zero except for a long-term (0.11 MHz) ringing as observed on the neutral ac data. This is not consistent with the 39-V Zener device protection, but does meet the criterion of 1 kV, 10 A. The response is as if a 150-V Zener device (not a spark gap) is actually in place. The ac coupler lines are hard to EMP and to half of all lightning strokes.

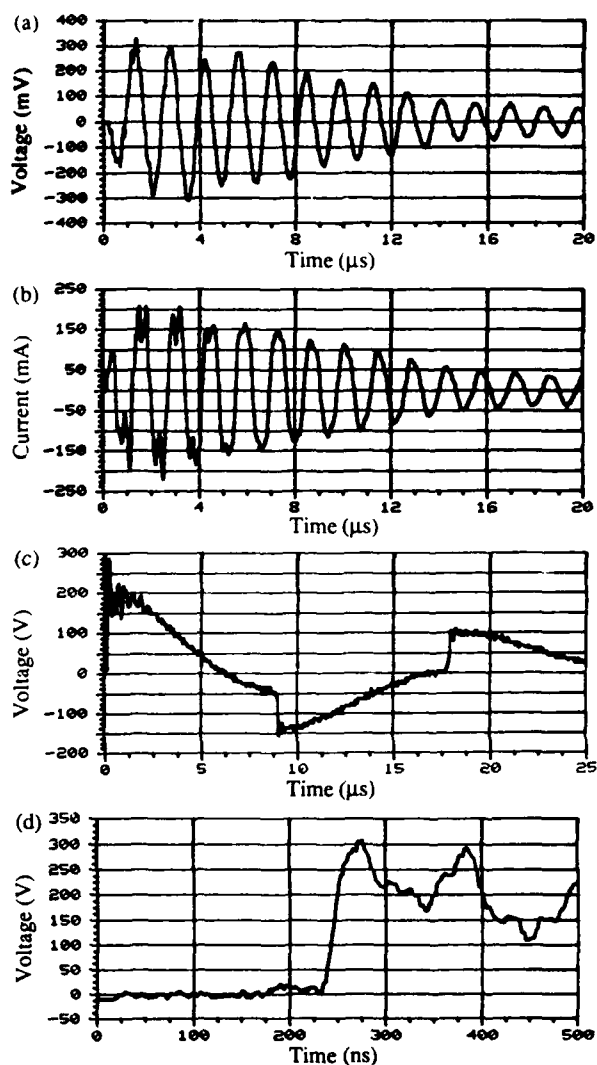


Figure 7. Response of ac power coupler line: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to injection (late time ringdown), and (d) voltage response to injection (early time peak).

3.7 Digital Signal Lines

The response of the digital lines to REPS (fig. 8) shows 4.5 V outside and 4 V inside the HFRF. This is less than the 12-V Zener protection. The current is similar in ringing but reads 75 mA (60 Ω) going into the HFRF and 19 mA (210 Ω) inside the HFRF. This factor of four seems large for the impedance mismatch from cable to shelter, even for a 0.67-MHz ringing frequency, but there is no filter or insertion loss involved. The ringing corresponds to the cable length. Extrapolated to threat levels, the inside values would correspond to 125 V and 0.6 A, which the 12-V Zener will clamp to an overshoot under 20 V. The injection response has a transient overshoot of 120 V, 380 mA (315 Ω), with a half-power duration of about 50 ns, reflecting a slow switching time. There may be substantial parasitic inductance to slow the Zener action down. The pulse continues around 30 V for several microseconds, as is expected. The bulk resistance of the MOS device is about 0.5 Ω , which accounts for the additional voltage as the large amount of current is shunted to ground. The 380-mA current shown in the data is what is left after the Zener current is drained off.

The digital signal lines are judged to be hard to HEMP and half of all lightning strokes.

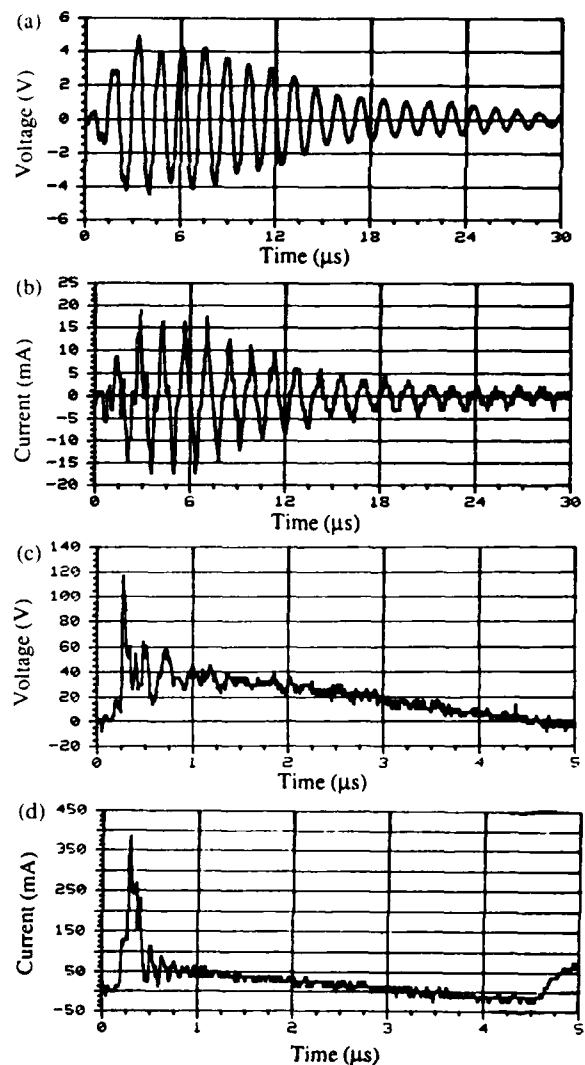


Figure 8. Response of digital signal lines: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to injection, and (d) current response to injection.

3.8 Audio Signal Lines

The response of the audio signal lines to the REPS environment (fig. 9) shows 210 mV, 1.5 mA (140 Ω), demonstrating the effectiveness of the cable shield and its termination. This response extrapolates to 6.6 V and 47 mA for threat level, well below the 12-V Zener protection provided by the 11845-10 device. The characteristic 0.67-MHz ringing for 300 ft is changed to a 0.53-MHz ringing because of the slower velocity of conduction within the shielded cable. The HFRF audio signal lines are demonstrated to be hard to HEMP even without the protective devices.

The lightning injection pulse is applied to the conductor within the shield in order to test the capacity of the protection devices. This is an extreme test since the shield would provide excellent protection, as was demonstrated by the REPS data. A 130-V, 0.5-A (260- Ω) response was observed, with the voltage duration being 70 ns. This indicates a slow switching time, but is not reason to doubt the hardness of the circuit.

The audio signal lines are judged to be hard to both HEMP and half of all lightning strokes.

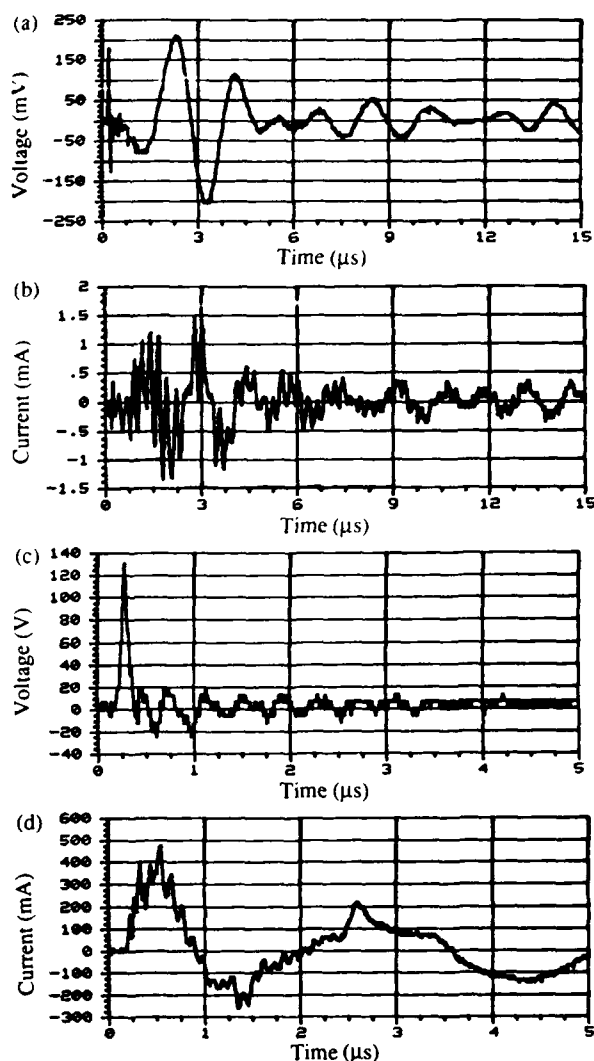


Figure 9. Response of audio signal lines: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to Injection, and (d) current response to Injection.

3.9 Telephone Lines

The response of the telephone signal lines to the REPS HEMP (fig. 10) shows 500 mV, 2.8 mA (180 Ω). When extrapolated to threat levels, these would register as 15.6 V and 87 mA. The 11845-100 protective device with a 100-V Zener would clamp these levels (as demonstrated by the injection test data) well within the criterion of 100 V, 10 A. These low levels are a result of the shielding protection around the lines. The ringing shows characteristics of the 300-ft line and a slower propagation velocity. The HFRF telephone lines are clearly hard to HEMP.

The lightning injection pulse was applied to the interface without cable attached, to test the capacity of the protective device. A surge of 180 V, 450 mA lasted for 50 ns, showing a slow switching time for the device. Levels promptly settled to zero, indicating a spark gap. Although the voltage criterion is exceeded, this circuit is probably hard to 50 percent of all lightning strikes, since the shield of the cable would divert much more than 10 times the energy injected. The HFRF telephone lines are judged hard to the average lightning threat.

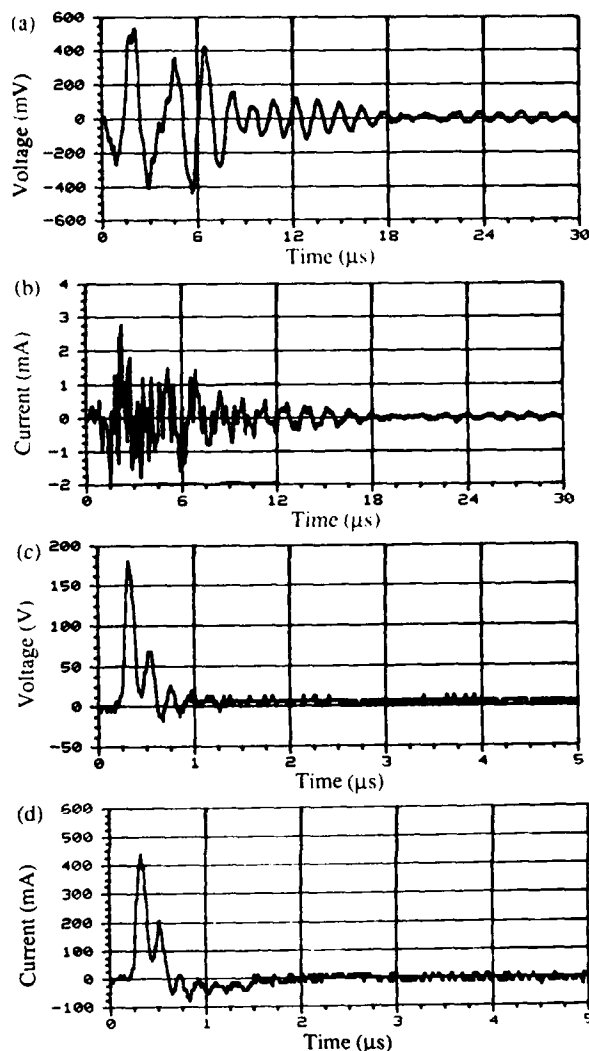


Figure 10. Response of telephone signal lines: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to injection, and (d) current response to injection.

3.10 Teletype Lines

The response of the teletype signal lines to the REPS HEMP (fig. 11) shows a peak voltage of 65 mV and a peak current of 20 mA (3.3 Ω). When extrapolated to threat levels, these register as 2 V and 650 mA. The 11895-20 protective device with a 22-V Zener would clamp these levels (as demonstrated by the injection test data) well within the criterion. These low levels are a result of the shielding protection around the lines. The characteristic ringing again shows the 300-ft line and a slower propagation velocity. The HFRF telephone lines are clearly hard to HEMP.

The lightning injection pulse was applied to the interface without cable attached in order to test the capacity of the protective device. A surge of 120 V, 370 mA (324 Ω) lasts for 50 ns, showing a slow switching time for the device. Levels promptly settle to zero, indicating a spark gap. Although the voltage criterion is exceeded, this circuit is judged hard to half of all lightning strikes since the shield would divert most of the energy from a lightning stroke.

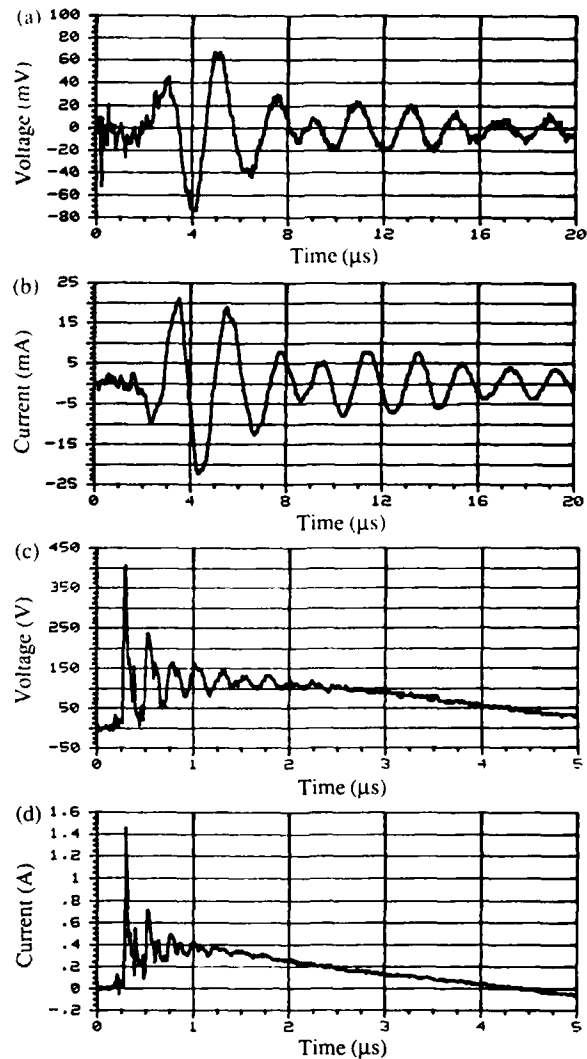


Figure 11. Response to teletype signal lines: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to injection, and (d) current response to injection.

3.11 Transmitter Lines

The transmitter protection device was connected between a tuned circuit load and a dipole antenna for the REPS tests. This dipole was on the centerline, 15 ft above the ground. The response of the transmitter mode to REPS HEMP (fig. 12) shows 225 V, 20 mA (1250 Ω , 50 Ω at tuned frequency only) on the center of the coax. When extrapolated to threat levels these would register as 7 kV, 0.5 A. The protective device with a 2-kV spark gap should reduce these levels to a level well within the criterion, but does so too slowly, as shown by the injection data. Since only 20 ns of transient will survive, the rf tuning elements can store the energy and attenuate the level safely for the next components. The 15-MHz ringing is due to the tuned frequency of the dipole and the tuned transmitter, dominating the frequency characteristic for the 270-ft length of coax. These lines are judged hard to HEMP.

The lightning injection pulse was applied to the interface without the cable attached in order to test the capacity of the device. A surge of 2.4 kV lasted for 20 ns, showing a faster switching time for the device. Arcing was observed inside the transmitter, but since lightning would strike the outside of the coax, which is grounded, this circuit is judged to be hard to an average lightning stroke.

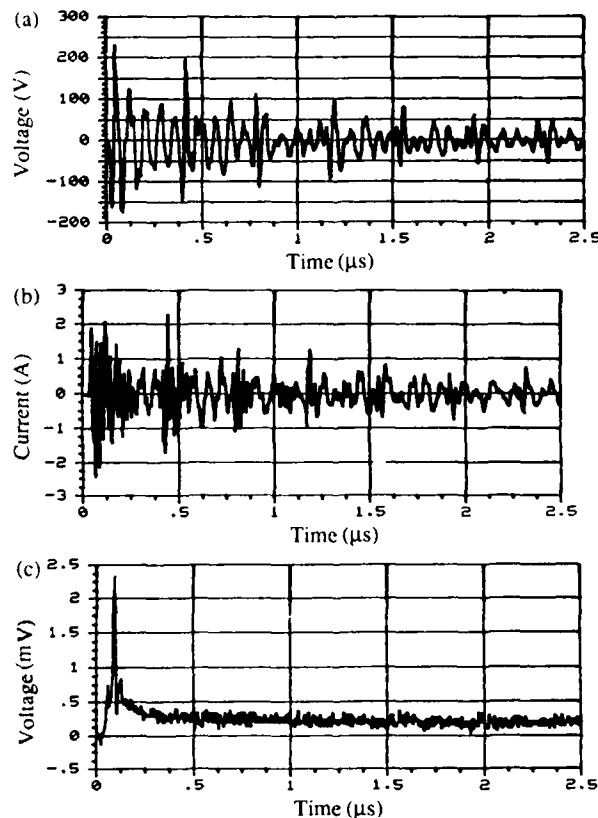


Figure 12. Response of transmitter lines: (a) voltage response to REPS, (b) current response to REPS, and (c) voltage response to injection.

3.12 Receiver rf Lines

The receiver 250-V protection device was connected between the 50- Ω load and the 2-kV protection device at the shield wall. Response of the receiver mode to REPS HEMP (fig. 13) shows 120 V, 3.5 A (34 Ω) on the center of the coax. Extrapolated to threat levels, these register as 3.8 kV and 110 A. The protective devices with a 2-kV spark gap and a 250-V Zener should reduce these levels to a level well within the criterion. The 15-MHz ringing is the tuned frequency of the dipole, but damps out faster than that for the transmitter because of the losses in the 50- Ω load compared to the tuned circuit. Like the transmitter, the receiver couples almost all of its energy from the antenna and very little energy from the coax. The HFRF rf lines are judged hard to HEMP.

The lightning injection pulse was applied to the interface without the cable attached. A surge of 2.8 kV lasted for 20 ns, showing the same switching time as for the transmitter. The voltage criterion is exceeded, but with the cable attached this circuit is judged hard to an average lightning stroke.

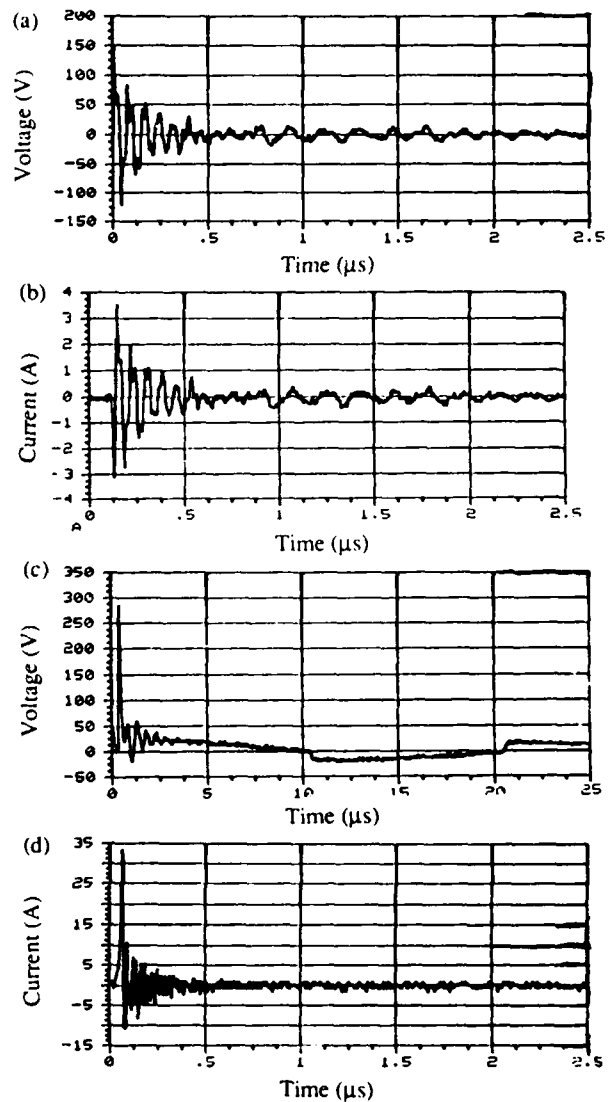


Figure 13. Response of receiver rf lines: (a) voltage response to REPS, (b) current response to REPS, (c) voltage response to Injection, and (d) current response to Injection.

4. CONCLUSIONS

The hf radio shield room in a configuration approved by the FAA was tested for EMP protection and was judged to be hardened to HEMP. The same configuration was tested for a hardness to half of all probable lightning strikes and was judged to be hard to this threat as well (see table 5).

Some test points failed the hardness criteria, but were passed, as explained when the data were presented. The field values from REPS are extrapolated and do not show protection device effects. Injection bypasses shielding for test points COU, DAT, AUD, TEL, KEY, XMT, and RCV.

There is reason to think with some confidence that the hf radio shield room, as configured, is hard to an HEMP environment and to half of all lightning strokes. The criteria notwithstanding, most (shield-circumvented) injected excess voltages and currents are of short duration (50 ns), and the parasitic reactances of the harness and shelter would reduce these substantially before the equipment

was exposed. The energy in these spikes is very small, and the endurance of components increases as the duration of stress decreases. Nonetheless, arcing in the XMT tuner did occur. A definitive answer to this issue would require a test with the equipment and wire harness in place and measurements made at the equipment. Possible insight may be gained by characterization tests on the protective devices to identify switching times, voltage overshoot from current and resistance, and possible installation reactances (circuit or grounding) which may be improved.

A serious problem was not addressed. The HFRF power protection design includes a gas-filled spark gap without self-quenching. As an EMP or lightning transient ionizes this gas, the ac power supply maintains the ionization until the resultant short circuit causes a circuit breaker to throw. This shuts down the radios, with the loss of any ongoing communications links. To avoid this upset it is recommended that a self-quenching type of spark gap be used to replace the existing one.

Table 5. HFRF HEMP/lightning test measurement points

Test point	Function	REPS scaled threat HEMP		Injected lightning		Criteria		Results	
		(V)	(mA)	(V)	(mA)	(kV)	(A)	By criteria	By judgement
SAC	Source power	320	16,000	400	14,000	1	10	fail	pass
AAC	Auxiliary power	375	16,000	(bad data)		1	10	fail	pass
COU	Coupler power	10	6,000	200	7,000	1	10	fail	pass
DAT	Digital line	125	600	120	380	0.1	10	pass	pass
AUD	Audio line	6.6	47	130	550	0.1	10	pass	pass
TEL	Telephone line	15.6	87	180	450	0.1	10	pass	pass
KEY	Teletype line	2	650	400	1,500	0.1	10	pass	pass
XMT	rf line transmit	7000	500	2400	na	10	500	pass	pass
RCV	rf line receive	3800	110,000	290	34,000	10	500	pass	pass

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